

# Solid State Physics p-n junctions

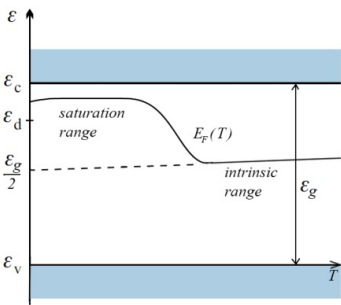
intrinsic semiconductor:  $E_F = \frac{1}{2}(\epsilon_v + \epsilon_c) + \frac{3}{4} k_B T \ln \frac{m_h}{m_c}$

doped (added a small portion of other atoms)  $\Rightarrow$  extrinsic semiconductor

$k_B T = 0.0258 \text{ eV}$  dopants impurities

$\epsilon_c - \epsilon^{(d)} \sim 0.001 \text{ eV}$   $\epsilon^{(a)}$   $\epsilon^{(a)}$   $\epsilon^{(a)} - \epsilon_v \sim 0.001 \text{ eV}$

$\Delta \epsilon^{(d)}$   
 $\Delta \epsilon^{(a)}$   $\} < 3 k_B T \Rightarrow$  extrinsic semic. are degenerate



$E_F^{(d)} = \frac{1}{2}(\epsilon_c + \epsilon^{(d)}) + \frac{3}{4} k_B T \ln \frac{m_h}{m_c}$

$E_F^{(a)} = \frac{1}{2}(\epsilon_v + \epsilon^{(a)}) + \frac{3}{4} k_B T \ln \frac{m_h}{m_c}$

$n_c^{(d)} = N_c e^{-(\epsilon_c - E_F^{(d)})/k_B T}$

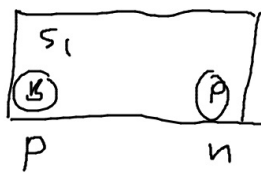
$p_v^{(a)} = P_v e^{-(E_F^{(a)} - \epsilon_v)/k_B T}$

$n_i = N_c e^{-(\epsilon_c - E_F)/k_B T}$

$p_i = P_v e^{-(E_F - \epsilon_v)/k_B T}$

$n_c^{(d)} = n_i e^{-(E_F^{(d)} - E_F)/k_B T}$

$p_v^{(a)} = p_i e^{-(E_F - E_F^{(a)})/k_B T}$



How to dope?

E.g. ion implantation then annealing

no movable charge carriers

$\Downarrow$  so called

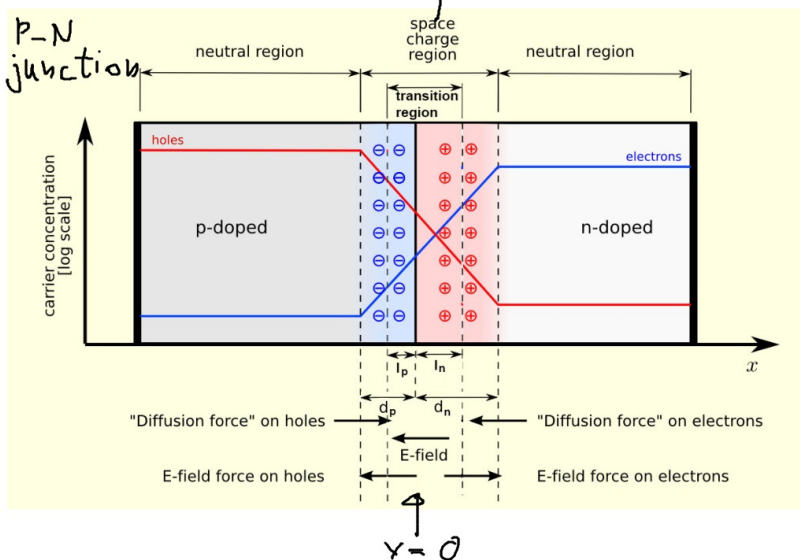
depletion region

creates an electric potential  $\psi(x)$ , which is the solution of Poisson's equation:

$\frac{\partial^2 \psi}{\partial x^2} + \frac{\partial^2 \psi}{\partial y^2} + \frac{\partial^2 \psi}{\partial z^2} = - \frac{\rho(x)}{\epsilon}$

charge density

$\epsilon = \epsilon_0 \cdot \epsilon_r$

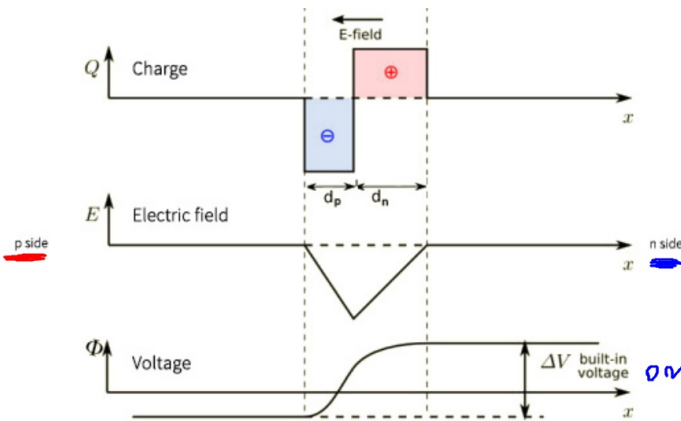


$$1D \Rightarrow \frac{d^2 \psi}{dx^2} = -\frac{\rho(x)}{\epsilon} \quad \rho(x) = e(N_d(x) - N_a(x) + p_v(x) - n_c(x))$$

$$\rho(x) = \begin{cases} 0 & x < -d_p \\ eN_a & -d_p < x < 0 \\ -eN_d & 0 < x < d_n \\ 0 & d_n < x \end{cases}$$

$$d_p = \sqrt{\frac{N_d \epsilon \psi_0}{N_a(N_a + N_d) 2e}} \quad d_n = \sqrt{\frac{N_a \epsilon \psi_0}{N_d(N_a + N_d) 2e}}$$

The un-biased ( $V_{ext} = 0$ ) p-n junction



$$n_c \cdot p_v = n_i^2$$

space-charge creates a Built-in potential  $\Delta \psi_0$

$$\Delta \psi_0 = \psi(\infty) - \psi(-\infty)$$

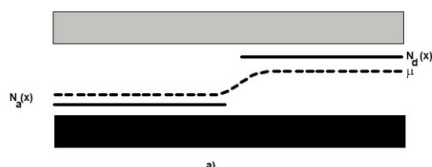
$$e \Delta \psi_0 = k_B T \ln \frac{N_d N_a}{n_i^2}$$

$$n_c(T, x) = N_c(T) e^{-(\epsilon_c - \epsilon_F(x))/k_B T}$$

$$p_v(T, x) = p_v(T) e^{-(\epsilon_F(x) - \epsilon_v)/k_B T}$$

$\epsilon_F$  (or  $\mu$ -chem. pot.) is a function of  $x$        $\mu(x) = \epsilon_F(x)$

Two different views of the same thing



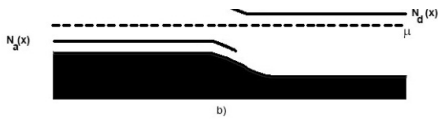
$\mu = \mu(x)$ , bands are straight

$\epsilon_v, \epsilon_c$  constant

(does not depend on  $x$ )

or

$\epsilon_F$  constant, the bands bend

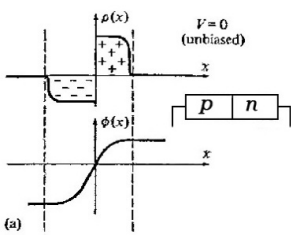


$\epsilon_c(x)$   
 $\epsilon_p = \mu$  const  
 $\epsilon_v(x)$

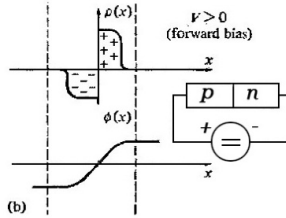
Biased p-n junctions  $V_{ext} \neq 0, E_{ext} \neq 0$

means  $V_{ext} \neq 0$

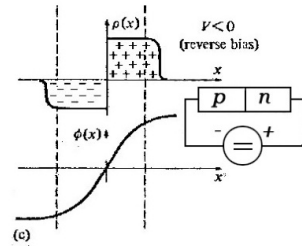
$V = \Delta\phi_0 - V_{ext}$



$V=0$



$V > 0$



$V < 0$

just consider the holes for the moment  
Two types of current:

- holes from p-side move through the depletion region to the n-side where they recombine w electrons  
 $j_h^{(rec)}$  recombination current  
 depends on  $V$  (or  $V_{ext}$ )

$j_h^{(rec)}(V) = C \cdot e^{-e\Delta\phi/k_B T}$

$\Delta E = e\Delta\phi$

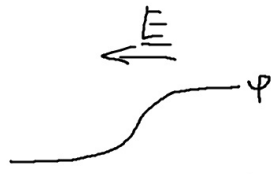
$\Delta\phi = \Delta\phi_0 - V_{ext} \Rightarrow j_h^{(rec)}(V) = C e^{-e\Delta\phi_0/k_B T} \cdot e^{eV_{ext}/k_B T}$

- holes generated on n-side (!) near the depl. region will move to the p-side ( $\underline{E}$  is from n  $\rightarrow$  p)

$j_h^{(gen)}$  generation current  
independent of  $V$  (or  $V_{ext}$ )

why?

$E = -\frac{\Delta\phi}{d}$



hole generated at the n-side if wanders into the depletion region is swept to the p-side by  $\underline{E}$   
 $\Rightarrow j_h^{gen}$  independent of  $V$

$v_h \sim 10^5$  u/s

$\langle v_h \rangle = v_{drift} \cdot j_h^{gen}$   
 $10^{-3}$  u/s

10-1000 nm

$j_h^{(gen)} = ?$

in equilibrium when  $V_{ext} = 0$

$$j_h^{(gen)} + j_h^{(rec)} = 0$$

$$j_h^{(gen)} = -j_h^{(rec)}(V_{ext}=0) = -C e^{-e\Delta\phi_0/k_B T}$$

The total current of holes:

$$j_{tot} = j_h^{(gen)} + j_h^{(rec)} = j_h^{(gen)} + \underbrace{C e^{-e\Delta\phi_0/k_B T}}_{= -j_h^{(gen)}} \cdot e^{eV_{ext}/k_B T}$$

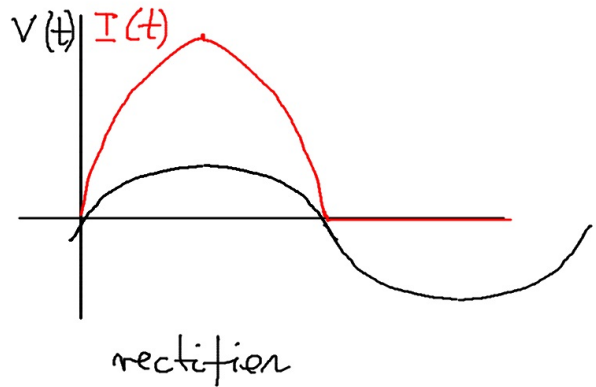
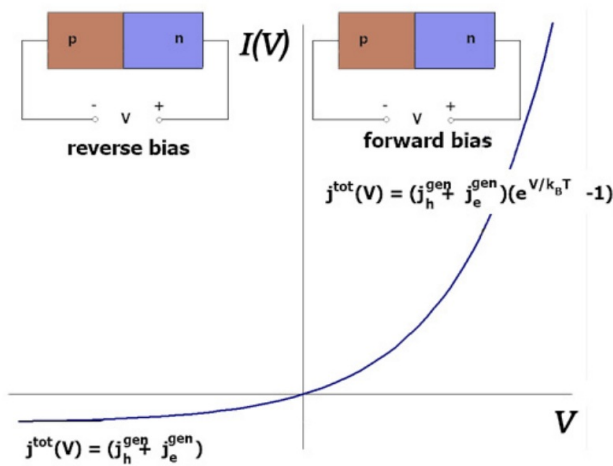
$$j_{tot}^{(h)} = j_h^{(gen)} (e^{eV_{ext}/k_B T} - 1)$$

Same for electrons

$$j_{tot} = j_{tot}^{(h)} + j_{tot}^{(e)} = \underbrace{(j_h^{(gen)} + j_e^{(gen)})}_{const} (e^{eV_{ext}/k_B T} - 1)$$

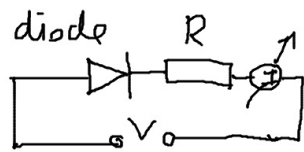
forward bias:  $V_{ext} > 0$

reverse -||-:  $V_{ext} < 0$



### 1. single p-n junction

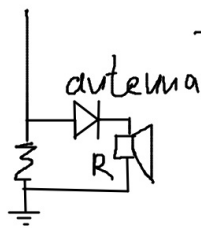
this device is the **diode**.



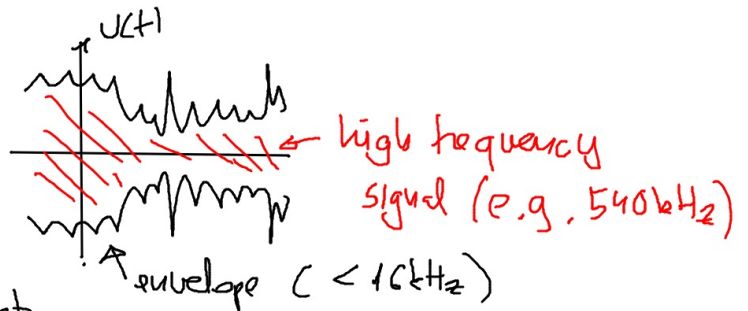
$$V = V_0 \sin \omega t \Rightarrow I(t) = \begin{cases} \frac{V_p}{R} \cdot \sin \omega t & , \text{when } V > 0 \\ 0 & V < 0 \end{cases}$$

Can be used for decoding AM (Amplitude modulated) radio waves:

near the transmitter

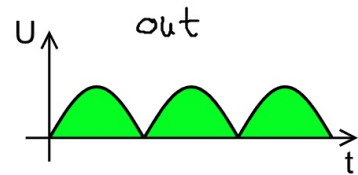
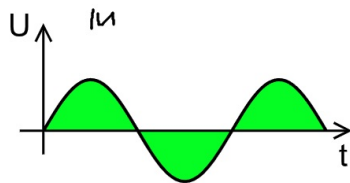
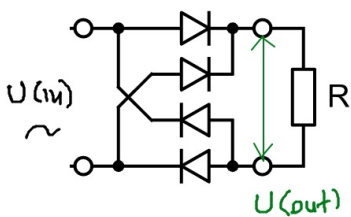


AM



the loudspeaker cannot follow the high frequency part, only the envelope  
diode: cuts out the negative part

Full wave rectifier (Graetz-bridge)



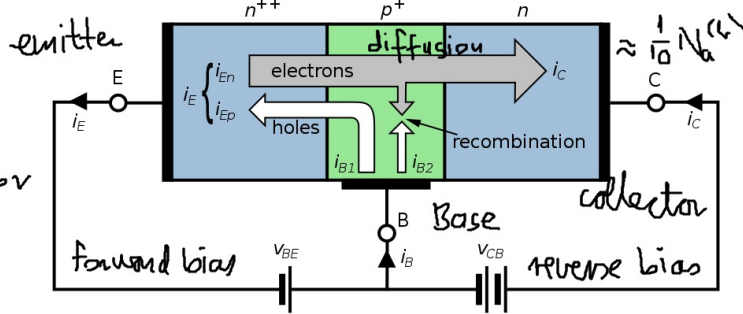
## 2. Bipolar Junction transistor (BJT)

1948

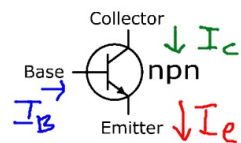
both electrons and holes

$$N_d \gg N_a \approx 10 N_d \quad \text{widths } w_e \gg w_b, w_c > w_e + w_b$$

n-p-n transistor



in circuits



narrow base layer  $\Rightarrow$  small recombination, most electrons ( $\approx 95\%$ ) passes through to the collector

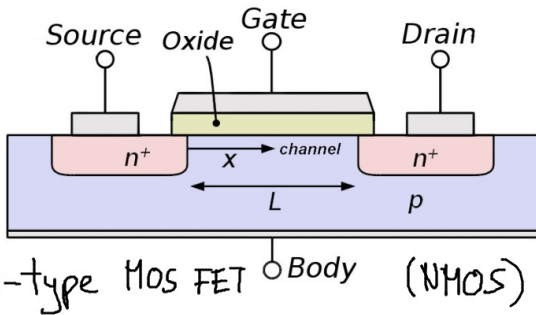
Two different mode of operation:

a)  $\Delta U_{BE} \rightarrow \Delta I_{BE} \rightarrow i_{EC} \gg i_{BE} \rightarrow \Delta U_{CE} \gg \Delta U_{BE} \Rightarrow$  amplifier

b) switch

If  $V_{BE}$  too is reverse biased  $\Rightarrow$  no current  $\Rightarrow$  used as switch  
 $V_{CB}$  too is forward bias  $\Rightarrow$  max current

field effect transistor - FET  $\leftarrow$  unipolar: either electrons or holes



n-type MOS FET (NMOS)  
 $V_{GS}$  controls a current through the p-type region in a channel between Source and Drain

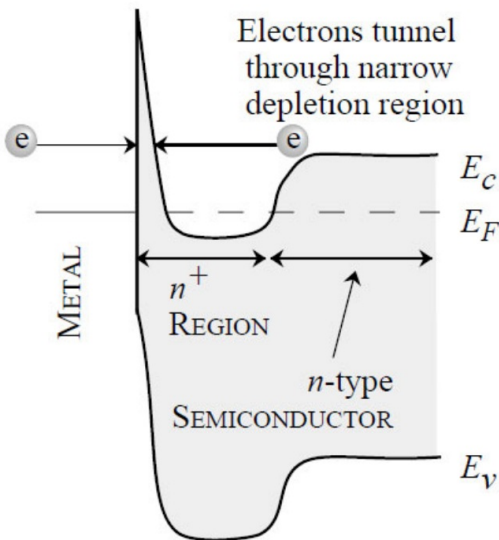
was created in 1945, was abandoned in 1950 (for BJT) used again from 1959 as MOSFET

advantages: smaller size (higher density)  
 lower power consumption  
 $\Rightarrow$  used in integrated circuits cell phones...

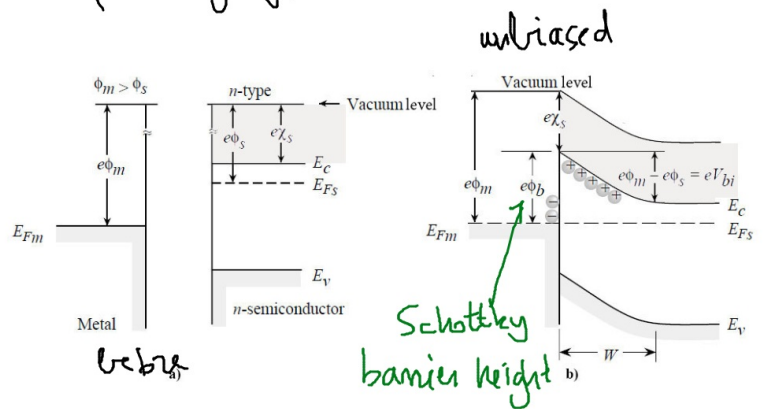
CMOS  
 Complementary MOS  
 $\Downarrow$   
 logic gate from NMOS + PMOS

### Metal-semiconductor junctions

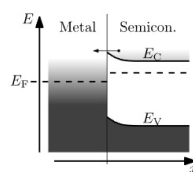
ohmic junctions



Schottky junction (rectifying)

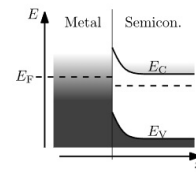


(for calculating the thermionic emission formula must be used)



forward

e can move into the metal



reverse

e can't move into the metal

bias